

The Diode Equation

The reason for calling the proportionality constant I_{sat} will become obvious when we consider reverse bias. Let us now make V_a **negative** instead of positive. The applied electric field now **adds in the same direction** to the built-in field. This means the barrier will **increase** instead of decrease, and so we have what is shown in [Figure 1](#). Note that we have marked the barrier height as $q(V_{\text{bi}} - V_a)$ as before. It is just that now, V_a is negative, and so the barrier is bigger.

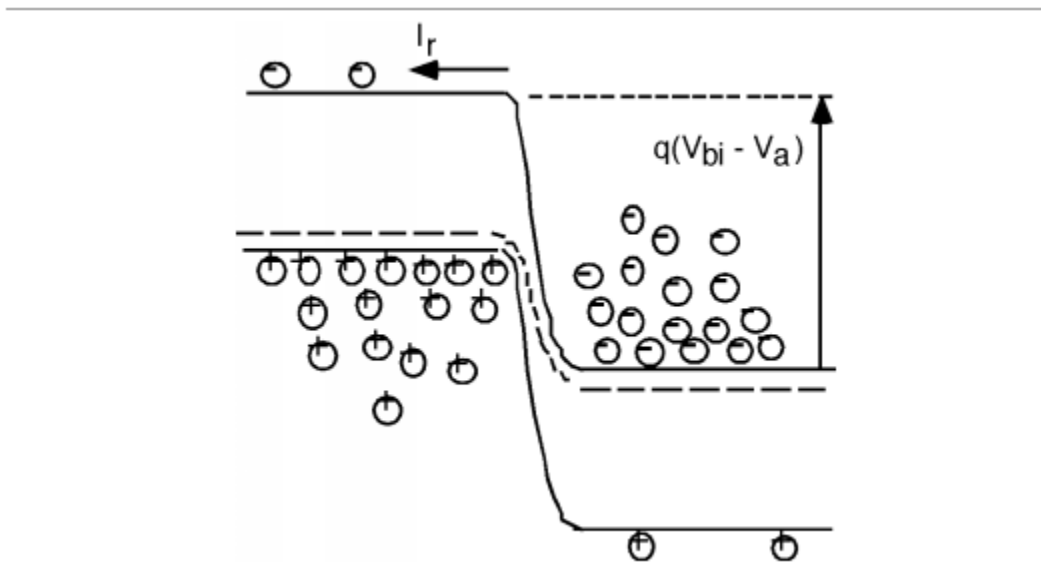


Figure 1: P-N junction under reverse bias ($V_a < 0$)

Remember, the electrons fall off exponentially as we move up in energy, so it does not take much of a shift of the bands before there are essentially **no** electrons on the n-side with enough energy to get over the barrier. This is reflected in the [diode equation](#) where, if we let V_a be a negative number, $e^{qV_a/kT}$ very quickly goes to zero and we are left with

$$I = -I_{\text{sat}}$$

(1)

Thus, while in the forward bias direction, the current increases exponentially with voltage, in the reverse direction it simply saturates at $-I_{\text{sat}}$. A plot of I as a function of voltage or an **I-V characteristic curve** might look something like [Figure 2](#).

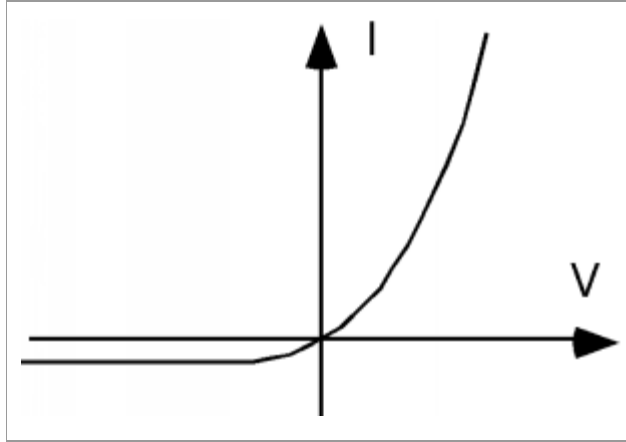


Figure 2: Idealized I-V curve for a p-n diode

In fact, for **real diodes** (ones made from silicon) I_{sat} is such a small value (on the order of 10⁻¹⁰ amps) that you can not even see it on most common measuring devices (oscilloscope, digital volt meter etc.) and if you were to look on a device called a **curve tracer** (which you will learn more about in Electronic Circuits [ELEC 342]) what you would really see would be something like [Figure 3](#).

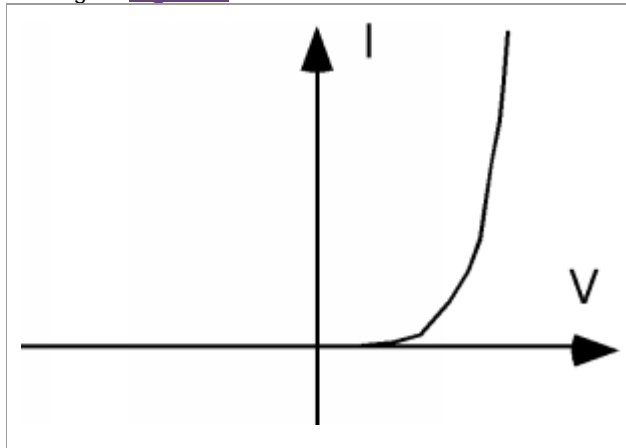


Figure 3: Realistic I-V curve

We see what looks like zero current in the reverse direction, and in fact, what appears to be no current until we get a certain amount of voltage across the diode, after which it very quickly "turns on" with a very rapidly increasing forward current. For silicon, this "turn on" voltage is about 0.6 to 0.7 volts.

Digital volt meters (DVM's) use this characteristic for their "diode check" function. What they do is, when the "red" or positive lead is connected to the p-side (anode, or arrow in the diagram) and the "black" or negative lead is connected to the n-side (cathode, or bar in the diagram) of a diode, the meter attempts to pass (usually) 1 mA of current through the diode. If the 1 mA of current is allowed to flow, the meter then indicates the amount of forward voltage developed across the diode. If it reads something like 0.673 volts, then you can be pretty sure the diode is

OK. Reverse the leads, and the diode is reverse biased, and the meter should read "OL" (overload) or something like that to indicate that no current is flowing.

The diode equation is usually approximated by two somewhat simpler equations, depending upon whether the diode is forward or reverse biased:

$$I \approx \begin{cases} 0 & \text{if } V_a < 0 \\ I_{\text{sat}} e^{qV_a/kT} & \text{if } V_a > 0 \end{cases}$$

(2)

For reverse bias, as we said, the current is essentially nil. In the forward bias case, the exponential term quickly gets much larger than unity, and so we can forget the "-1" term in the diode equation. Remember, we said that kT at room temperature had a value of about 1/40 of an eV, so $qkT \approx 40V^{-1}$, this means we can also say for forward bias that

$$I = I_{\text{sat}} e^{40V_a}$$

(3)

From this equation it is easy to see that only a small positive value for V_a is needed in order to make the exponential much greater than unity.

Now let's connect this "ideal diode equation" to the real world. One thing you might ask yourself is "How could I check to see if an actual diode follows the equation given here?" As we said, I_{sat} is a very small current, and so trying to do the reverse test is probably not going to be successful. What is usually done is to measure the diode current (and forward voltage) over several orders of magnitude of current.

NOTE:

While the current can vary by many orders of magnitude, the voltage is more or less limited to values between 0 and 0.6 to 0.7 volts, not by any fundamental process, but rather simply by the fact that too much forward current will burn up the diode.

If we take the natural log of both sides of the second piece of Equation 2, we find:

$$\ln(I) = \ln(I_{\text{sat}}) + qV_a/kT$$

(4)

Thus, a plot of $\ln(I)$ as a function of V_a should yield a straight line with a slope of qkT , or 40.

Well, I went into the lab, grabbed a real diode and made some measurements. Figure 4 is a plot of the natural log of the current as a function of voltage from 0.05 to 0.70 volts. Included with this plot, is a linear curve fit to the data which is plotted as a dotted line. The linear fit goes through the data points quite nicely, so the current is surely an exponential function of the applied voltage! From the expression for the best fit, which is printed above the graph,

we see that $\ln(I_{\text{sat}}) = -19.68$. That means that $I_{\text{sat}} = e^{-19.68} = 2.89 \times 10^{-9}$ amps, which is indeed a very small current. Look at the slope however. Its supposed to be 40, and yet it turns out to be slightly more than 20! This comes about because of some complex details of exactly what happens to the electrons and holes when they cross the junction. In what is called the **diffusion dominated situation** electrons and holes are injected across the junction, after which they diffuse away from the junction, and also recombine, until eventually they are all gone. This is shown schematically in [Figure 5](#). The other regime is called **recombination dominated** and here, the majority of the current is made up of the electrons and holes recombining directly with each other at the junction. This is shown in [Figure 6](#). For recombination dominated diode behavior, it turns out that the current is given by

$$I = I_{\text{sat}} e^{qV_a/2kT}$$

(5)

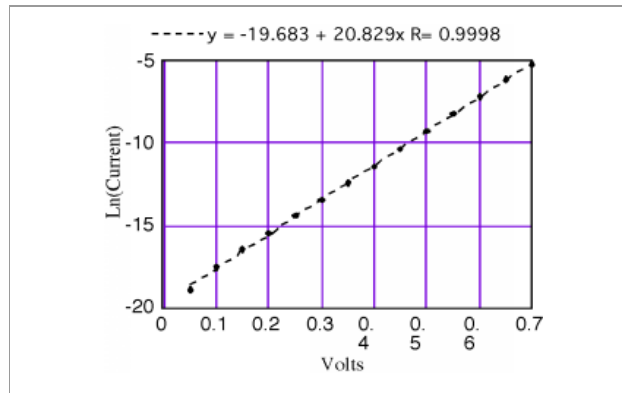


Figure 4: Plot showing $\ln(I)$ as a function of V_a for a 1N4123 silicon diode

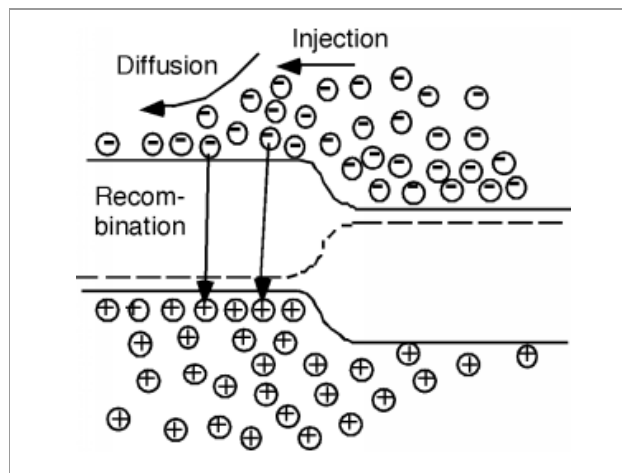


Figure 5: Diffusion dominated diode behavior

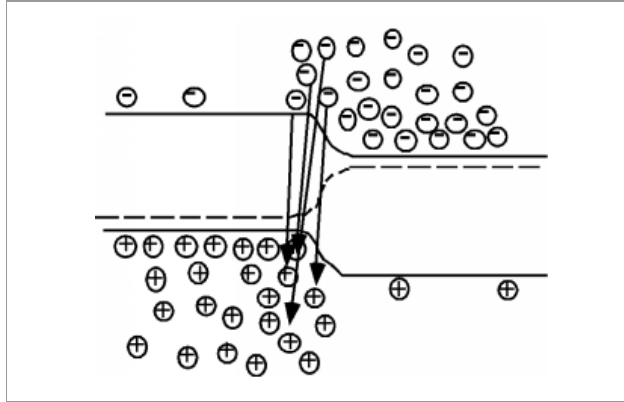


Figure 6: Recombination dominated diode behavior

In general, a particular diode might have a combination of these two effects going on, and so people often use a more general form for the diode equation:

$$I = I_{\text{sat}} e^{qV/nkT}$$

(6)

where n is called the **ideality factor** and is a number somewhere between 1 and 2. For the diode which gave the data for our example $n=1.92$ and so most of the current is dominated by recombination of electrons and holes in the depletion region.

Source: <http://cnx.org/content/m1008/latest/?collection=col10114/latest>